

## FM Receiver

### Description

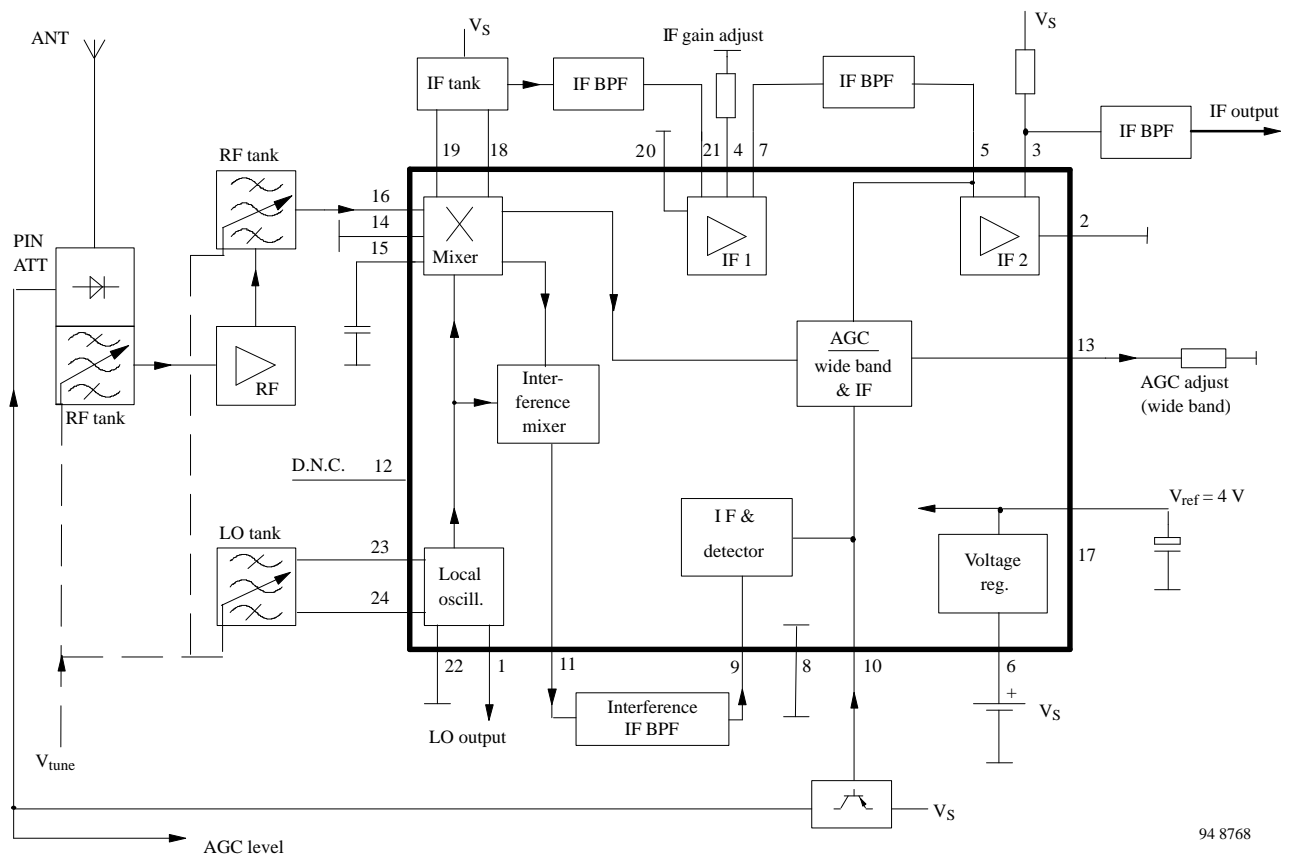
The IC U4065B is a bipolar integrated FM-frontend circuit. It contains a mixer, an oscillator, two IF preamplifiers and an unique interference sensor. The device is designed

for high performance car radio and home receiver applications.

### Features

- All frontend functions of a high performance FM-receiver, except the RF preamplifier, are integrated
- Improved dynamic range by high current double balanced mixer design and a new AGC conception with 3 loops on chip
- Improved blocking and intermod behavior by use of an unique "interference" sensor controlling the AGC
- Easy cascading of three IF filters (ceramic) by use of two on-chip IF preamplifiers
- On-chip control functions are available for system gain adjust (dB linear vs. dc current)
- Low noise LO design
- ESD protected

### Block Diagram



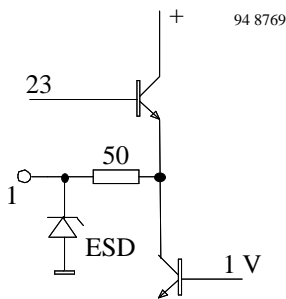
## U4065B

### Pin Description

| Pin | Symbol   | Function                                 |
|-----|----------|--|
| 1   | LOBUFF   | Buffered local oscillator output         |
| 2   | GND1     | Ground of the second IF amplifier        |
| 3   | IF2OUT   | Output of the second IF amplifier        |
| 4   | GAINIF1  | Gain control of the first IF amplifier   |
| 5   | IF2IN    | Input of the second IF amplifier         |
| 6   | VS       | Supply voltage                           |
| 7   | IF1OUT   | Output of the first IF amplifier         |
| 8   | GND2     | Ground                                   |
| 9   | IMIFIN   | Input of the amplifier for the IM-sensor |
| 10  | AGCOUT   | Output of the automatic gain control     |
| 11  | IMMIXOUT | Output of the intermodulation mixer      |
| 12  | D.N.C.   | Do not connect                           |

| Pin | Symbol  | Function                                 |
|-----|---------|--|
| 13  | AGCWB   | Threshold adjustment of the wideband AGC |
| 14  | GND3    | Mixer ground                             |
| 15  | MIXIN1  | Input 1 of the double balanced mixer     |
| 16  | MIXIN2  | Input 2 of the double balanced mixer     |
| 17  | VREF    | Reference voltage output                 |
| 18  | MIXOUT1 | Mixer output 1                           |
| 19  | MIXOUT2 | Mixer output 2                           |
| 20  | GND4    | Ground of the first IF amplifier         |
| 21  | IF1IN   | Input of the first amplifier             |
| 22  | GND5    | Oscillator ground                        |
| 23  | LOE     | Local oscillator (emitter)               |
| 24  | LOB     | Local oscillator (base)                  |

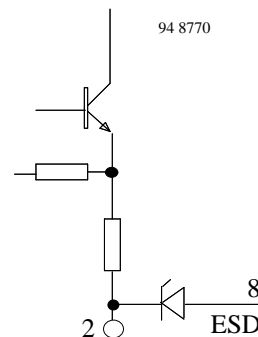
#### LOBUFF



#### Buffered local oscillator output:

It drives the FM-input of the PLL circuit (for example U428xBM-family). The typical parallel output resistance at 100 MHz is 70  $\Omega$ , the parallel output capacitance is about 10 pF. When using an external load of 500  $\Omega$  / 10 pF, the oscillator swing is about 100 mV. The second harmonic of the oscillator frequency is less than -15 dBc.

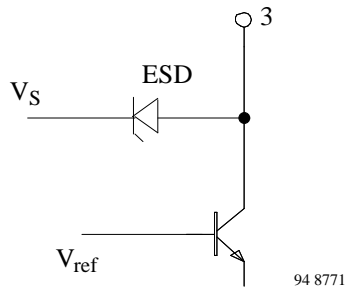
#### GND1



#### Ground of the second IF amplifier:

There is no internal connection to the other ground pins.

## IF2OUT

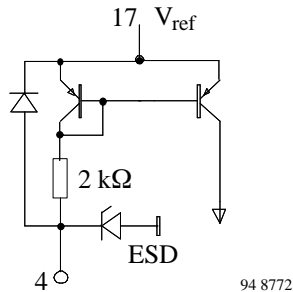


### Output of the second IF amplifier:

The parallel output capacitance to ground is about 7 pF. The external load resistance is to connect to  $V_S$ . The dc current into the pin is typically 3 mA.

**Note:** Supply voltage  $V_S$  has to be protected against IF-distortion

## GAINIF1



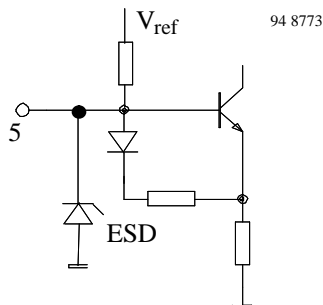
### Gain control of the first IF amplifier:

The gain of the first IF amplifier can be adjusted by a resistor to ground. This is useful for example to compensate the insertion loss tolerances of the ceramic BPF's. Please ensure that the output current of the pin does not exceed 150  $\mu$ A in any case. Linear increasing in the current out of GAINIF1 effects dB linear increasing of the gain (0.15 dB/ $\mu$ A).

$$I_4 = 0 \Rightarrow G = G_{\min} = 2 \text{ dB}$$

$$I_4 = 140 \mu\text{A} \Rightarrow G = G_{\max} = 22 \text{ dB}$$

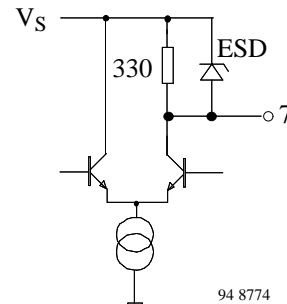
## IF2IN



### Input of the second IF amplifier:

The parallel input resistance is 330  $\Omega$ . The parallel input capacitance is about 12 pF. No dc current is allowed. To avoid overload of this stage an internal detector watches the input level and causes current at the AGCOUT pin.

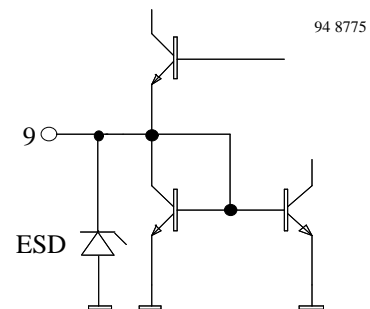
## IF1OUT



### Output of the first IF amplifier:

The parallel output resistance is 330  $\Omega$  which allows the use of a standard ceramic BPF. The parallel output capacitance is about 7 pF. The dc voltage at the pin is 0.5 V less than  $V_S$ .

## IMIFIN

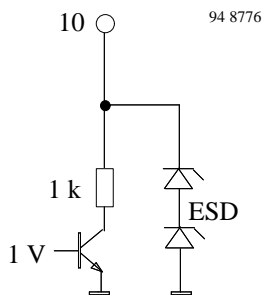


### Input of the IF amplifier for the IM-sensor:

The parallel input resistance is 330  $\Omega$ . The amplifier is extremely sensitive to ac signals. A few hundred  $\mu$ V of IF-signal at this pin will cause current at the AGC output. Therefore pay attention when connecting the standard ceramic filter used between IMOUT and this pin. The reference point of the filter has to be free of any ac signal. Please avoid dc current at this pin.

## U4065B

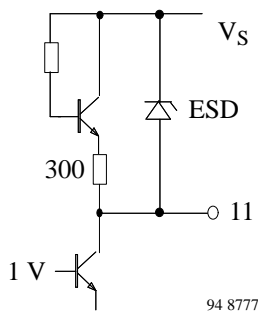
### AGCOUT



#### Output of the automatic gain control:

The AGC output is an open collector output. The current of the pin diode is this current multiplied by the current gain of the external PNP transistor. The dc voltage at the pin may vary from 2 V to  $V_S$ , therefore you can easily use this pin as an indicator of the AGC regulation state.

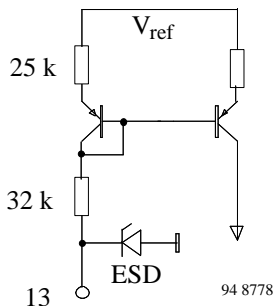
### IMMIXOUT



#### Output of the intermodulation mixer:

The parallel output resistance is 330  $\Omega$  which allows the use of a standard ceramic BPF without any further matching network. Please ensure that the ground-pin of the filter is free of ac signals.

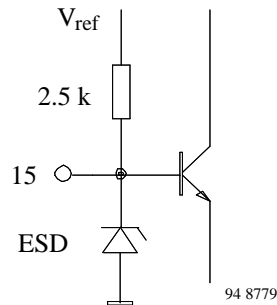
### AGCWB



#### Threshold adjustment of the wideband AGC:

The threshold of the wideband AGC can be adjusted by an external resistor to ground. The setting range is 10 dB. For minimum blocking this pin is connected to ground. In order to set the threshold to smaller levels the resistance value should be up to a few hundred k $\Omega$ .

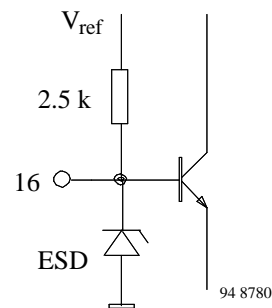
### MIXIN1



#### Input 1 of the double balanced mixer:

The parallel input resistance is 1.2 k $\Omega$ . The parallel input capacitance is about 9 pF. When using the mixer unbalanced this pin is to be grounded for RF-signals by an external capacitance of a few nF. DC current is not allowed.

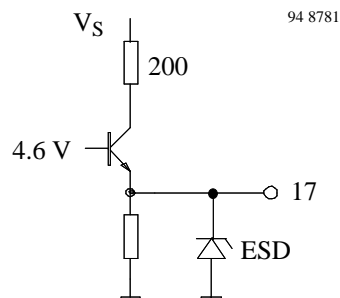
### MIXIN2



#### Input 2 of the double balanced mixer:

The parallel input resistance is 1.6 k $\Omega$ . The parallel input capacitance is about 7 pF. The double sideband noise figure of the unbalanced mixer is about 7 dB. In the balanced case the noise figure will be reduced by about 0.8 dB.

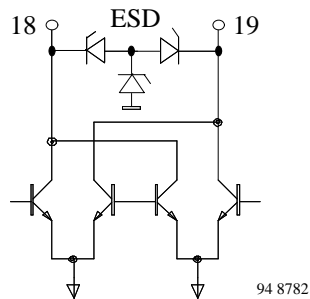
### VREF



#### Reference voltage:

The internal temperature compensated reference voltage is 3.9 V. It is used as bias voltage for most blocks, so the electrical characteristics of the U4065B are widely independent of the supply voltage. The internal output resistance of the reference voltage is less than 10  $\Omega$ . To avoid internal coupling across this pin external capacitors are required. The maximum output current is  $I_{ref} = 5$  mA.

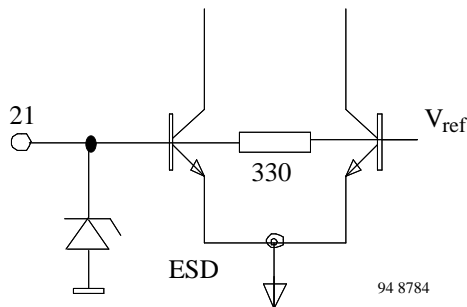
## MIXOUT1, MIXOUT2



### Mixer output 1, 2:

The mixer output is an open collector of a bipolar transistor. The minimum voltage at this pins is 5 V ( $V_S$ -voltage swing). The dc current into this pins is typically 9 mA. Good LO- and RF suppression at the mixer output can be achieved by symmetrical load conditions at the pins MIXOUT1 and MIXOUT2.

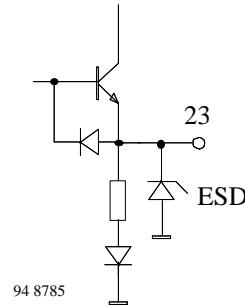
## IF1IN



### Input of the first IF amplifier:

The typical input resistance is 330  $\Omega$ . The dc voltage is nearly the same one as the reference voltage. Please avoid dc current at this pin.

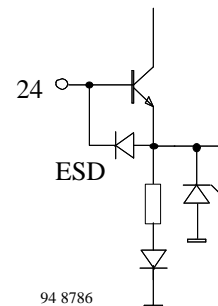
## LOE



### Emitter of the local oscillator:

An external capacitor is connected between LOE and ground. The ground pin of this capacitor is to connect to the pin GND5. GND5 is the chip internal ground of the local oscillator.

## LOB



### Base of the local oscillator:

The tank of the local oscillator is connected at pin LOB. The ground pin of this tank is to connect to the pin GND5. GND5 is the chip internal ground into pin 24 of the local oscillator. The resonant resistance of the tank should be about 250  $\Omega$ . Minimum Q of the unloaded tank is 50.

## U4065B

### Functional Description

The U4065B FM-frontend IC is the dedicated solution for high end car radios. A new design philosophy enables to build up tuners with superior behavior. This philosophy is based on the fact that the sensitivity of state of the art designs is at the physical border and cannot be enhanced any more. On the other hand, the spectral power density in the FM-band increases. An improvement of reception can only be achieved by increasing the dynamic range of the receiver. This description is to give the designer an introduction to get familiar with this new product and its philosophy.

#### 1. The Signal Path

The U4065B offers the complete signal path of an FM-frontend including a highly linear mixer and two IF preamplifiers. The mixer is a double balanced high current Gilbert Cell. A high transit frequency of the internal transistors enables the use of the emitter grounded circuit with its favorable noise behavior. The full balanced output offers LO carrier reduction.

The following IF preamplifier has a dB-linear gain adjustment by dc means. Thus different ceramic filter losses can be compensated and the overall tuner gain can be adapted to the individual requirements. The low noise design suppresses post stage noise in the signal path. Input- and output resistance is 330  $\Omega$  to support standard ceramic filters. This was achieved without feedback, which would cause different input impedances when varying the output impedance.

The second IF preamplifier enables the use of three ceramic filters with real 330  $\Omega$  input- and output termination. Feedthrough of signals is kept low. The high level of output compression is necessary to keep up a high dynamic range.

Beneath the signal path the local oscillator part and the AGC signal generation can be found on chip. The local oscillator uses the collector grounded colpitts type. A low phase noise is achieved with this access. A mutual coupling in the oscillator coil is not necessary.

#### 2. The AGC Concept

Special care was taken to design a unique AGC concept. It offers 3 AGC loops for different kinds of reception conditions. The most important loop is the interference sensor part.

In today's high end car radios, the FM AGC is state of the art. It is necessary to reduce the influence of 3rd and higher order intermodulation to sustain reception in the presence of strong signals in the band. On one hand, it makes a sense to reduce the desired signal level by AGC as few as possible to keep up stereo reception, on the other hand two or more

strong out of channel signals may interfere and generate an intermodulation signal on the desired frequency. By introducing input attenuation, the level of the intermod signal decreases by a higher order, whereas the level of the desired signal shows only a linear dependency on the input attenuation. Therefore input attenuation by pin diodes may keep up reception in the presence of strong signals.

The standard solution to generate the pin diode current is to pick up the RF-signal in front of the mixer. Because the bandwidth at that point is about 1.5 MHz, this is called wideband AGC. The threshold of AGC start is a critical parameter. A low threshold does not allow any intermodulation but has the disadvantage of blocking if there is only one strong station on the band or if the intermod signals do not cover the desired channel. A higher AGC threshold may tolerate a certain ground floor of intermodulation. This avoids blocking, but it has the disadvantage, that no reception is possible, if the interfering signals do generate an intermod signal inside the desired channel. This contradiction could not be overcome in the past.

With the new U4065B IC, a unique access to this problem appears. This product has an interference sensor on chip. Thus an input signal attenuation is only performed, if the interfering signals do generate an intermod signal inside the desired channel. If they do not, the still existing wideband AGC is yet active but at up to 20 dB higher levels. The optimum AGC state is always generated.

The figures 1 to 4 illustrate the situation. In figure 1 the AGC threshold of a standard tuner is high to avoid blocking. But then the intermod signal suppresses the desired signal. The interference sensor of the U4065B takes care that in this case the AGC threshold is kept low as illustrated in figure 2.

In figure 3 the situation is vice versa. The AGC threshold of a standard tuner is kept low to avoid intermod problems. But then blocking makes the desired signal level drop below the necessary stereo level. In this case, the higher wideband AGC level of the U4065B enables perfect stereo reception.

By principle, this interference sensor is an element with a third order characteristic. For input levels of zero, the output level is zero, too. With increasing input level, the output level is increased with the power of three, thus preferring intermod signals compared to linear signals. At the same time, a down conversion to the IF level of 10.7 MHz is performed. If a corresponding 10.7 MHz IF filter selects the intermod signals, an output is only generated, if an intermod signal inside the 10.7 MHz channel is present.

The circuit blocks interference sensor and IF & detector build up a second IF chain. In an FM system, the max deviation of a 3rd order intermod signal is the triple max deviation of the desired signal. Therefore the ceramic IF BPF between Pin 11 and Pin 9 may be a large bandwidth type. This external part is the only additional amount for

this unique feature.

A further narrow band AGC avoids overriding the second IF amplifier. The amplitude information of the channel is not compressed in order to maintain multipath detection in the IF part of the receiver.

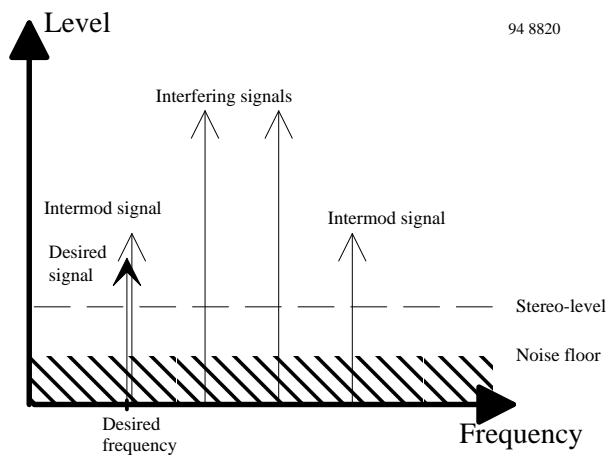


Figure 1 A high AGC threshold causes the intermod signal to suppress the desired signal

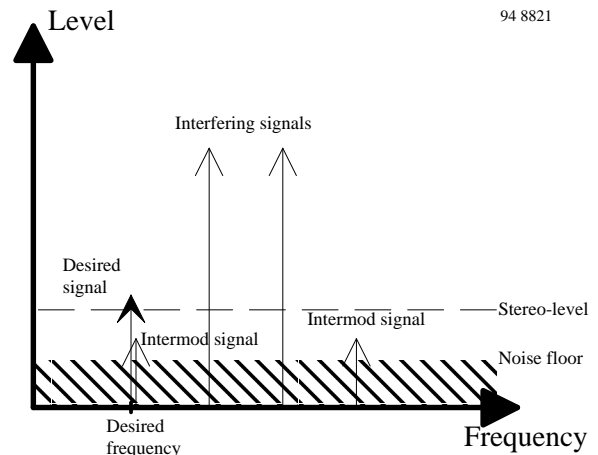


Figure 2 The correct AGC threshold of the U4065B provides optimum reception

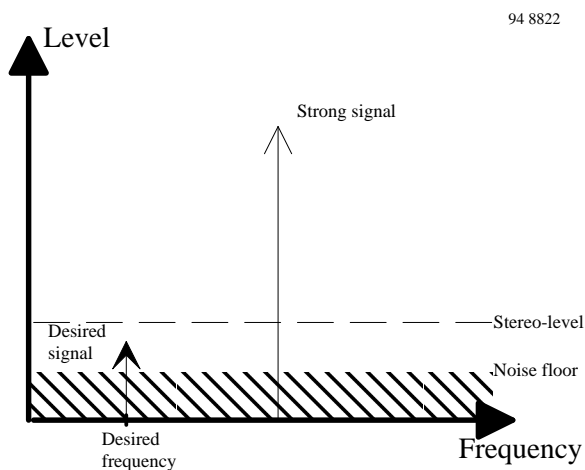


Figure 3 A low AGC threshold causes the blocking signal to suppress the desired signal

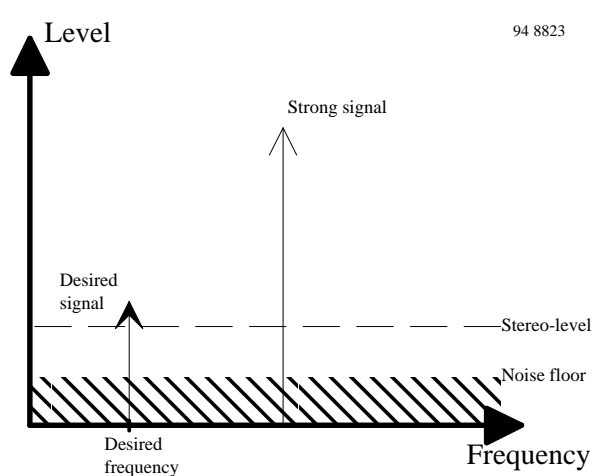


Figure 4 The correct AGC threshold of the U4065B provides optimum reception

### Absolute Maximum Ratings

Reference point is ground (Pins 2, 8, 14, 20 and 22)

| Parameters   | Symbol        | Value         | Unit             |
|--|---------------|---------------|------------------|
| Supply voltage   | $V_S$         | 10            | V                |
| Power dissipation at $T_{amb} = 85^\circ\text{C}$  | $P_{tot}$     | 470           | mW               |
| Junction temperature   | $T_j$         | 125           | $^\circ\text{C}$ |
| Ambient temperature range  | $T_{amb}$     | - 30 to + 85  | $^\circ\text{C}$ |
| Storage temperature range  | $T_{stg}$     | - 50 to + 125 | $^\circ\text{C}$ |
| Electrostatic handling:<br>Human body model (HBM),<br>all I/O pins tested against the supply pins. | $\pm V_{ESD}$ | 2000          | V                |

### Thermal Resistance

| Parameters         | Symbol     | Maximum | Unit |
|--------------------|------------|---------|------|
| Thermal resistance | $R_{thJA}$ | 90      | K/W  |

### Electrical Characteristics

$V_S = 8.0\text{ V}$ ,  $f_{RF} = 98\text{ MHz}$ ,  $f_{OSC} \cong 108.7\text{ MHz}$ ,  $f_{IF} = f_{OSC} - f_{RF} = 10.7\text{ MHz}$

Reference point ground (Pins 2, 8, 14, 20 and 22),  $T_{amb} = 25^\circ\text{C}$ , unless otherwise specified

| Parameters   | Test Conditions / Pins   | Symbol          | Min. | Typ. | Max. | Unit       |
|--|--|-----------------|------|------|------|------------|
| Supply voltage   | Pins 3, 6, 10, 18 and 19   | $V_S$           | 7    | 8    | 10   | V          |
| Supply current   | Pins 3+6+10+18+19  | $I_{tot}$       |      | 37   | 47   | mA         |
| <b>Oscillator</b> (GND5 has to be connected to external oscillator components) |  |                 |      |      |      |            |
| Oscillator voltage   | $R_{g24} = 220\ \Omega$ , unloaded Q<br>of $L_{OSC} = 70$ , $R_{L1} = 520\ \Omega$ |                 |      |      |      |            |
|  | Pin 24   | $V_{LOB}$       |      | 160  |      | mV         |
|  | Pin 23   | $V_{LOE}$       |      | 100  |      |            |
| Pin 1  | $V_{LOBUFF}$   |                 | 90   |      |      |            |
| Harmonics  | Pin 1  |                 |      |      | -15  | dBc        |
| Output resistance  | Pin 1  | $R_{LO}$        |      | 70   |      | $\Omega$   |
| Voltage gain   | Between pins 1 and 23  |                 |      | 0.9  |      |            |
| <b>Mixer</b> (GND3 has to be separated from GND1, GND2 and GND4)               |  |                 |      |      |      |            |
| Conversion power gain  | Source impedance:  | $G_C$           |      | 7    |      | dB         |
| 3rd order input intercept  | $R_{G15,16} = 200\ \Omega$   | $IP_3$          |      | 6    |      | dBm        |
| Conversion transconductance  | Load impedance:<br>$R_{L18,19} = 200\ \Omega$                                      | $g_C$           |      | 8    |      | mA/V       |
| Noise figure   |  | $NF_{DSB}$      |      | 7    |      | dB         |
| Input resistance to ground   | Pin 15   | $R_{ignd15}$    |      | 1.2  |      | k $\Omega$ |
| Input capacitance to ground  | $f = 100\text{ MHz}$   | $C_{ignd15}$    |      | 9    |      | pF         |
| Input resistance to ground   | Pin 16   | $R_{ignd16}$    |      | 1.6  |      | k $\Omega$ |
| Input capacitance to ground  | $f = 100\text{ MHz}$   | $C_{ignd16}$    |      | 7    |      | pF         |
| Input-input resistance   | Between Pin 15 and Pin 16  | $R_{ii15,16}$   |      | 1.6  |      | k $\Omega$ |
| Input-input capacitance  | Between Pin 15 and Pin 16  | $C_{ii15,16}$   |      | 5    |      | pF         |
| Output capacitance to GND  | Pin 18 and Pin 19  | $C_{ignd18,19}$ |      | 9    |      | pF         |



| Parameters   | Test Conditions / Pins   | Symbol                                 | Min. | Typ.           | Max. | Unit        |
|--|--|--|------|----------------|------|-------------|
| <b>First IF preamplifier (IF 1)</b>  |  |  |      |                |      |             |
| Gain control deviation by $I_4$  | Pin 4  |  |      | 20             |      | dB          |
| Gain control slope   |  | $dG_{IF1}/dI_4$                        |      | 0.15           |      | dB/ $\mu$ A |
| External control current to ground<br>at $G_{min}$<br>at $G_{nom}$<br>at $G_{max}$ |  | $I_{4min}$<br>$I_{4nom}$<br>$I_{4max}$ |      | 0<br>70<br>140 |      | $\mu$ A     |
| Power gain<br>at $I_{4min}$<br>at $I_{4nom}$<br>at $I_{4max}$                      | Between pins 21 and 7<br><br>Source impedance:<br>$R_{G21} = 200 \Omega$<br>Load impedance:<br>$R_{L7} = 200 \Omega$ | $G_{min}$<br>$G_{nom}$<br>$G_{max}$    |      | 2<br>12<br>22  |      | dB          |
| Noise figure<br>at $G_{max}$<br>at $G_{nom}$<br>at $G_{min}$                       |  | $NF_{min}$<br>$NF_{nom}$<br>$NF_{max}$ |      | 7<br>9<br>15   |      | dB          |
| Temperature coefficient of the gain at $G_{nom}$                                   |  | $TK_{nom}$                             |      | +0.045         |      | dB/K        |
| 1 dB compression at $G_{nom}$  | Pin 7  | $V_{cnom}$                             |      | 70             |      | mV          |
| -3 dB cutoff freq. at $G_{nom}$  | Pin 7  | $f_{cnom}$                             |      | 50             |      | MHz         |
| Input resistance   | Pin 21   | $R_{iIF1}$                             |      | 330            |      | $\Omega$    |
| Input capacitance  | $f = 10$ MHz   | $C_{iIF1}$                             |      | 5              |      | pF          |
| Output resistance  | Pin 7  | $R_{oIF1}$                             |      | 330            |      | $\Omega$    |
| Output capacitance   | $f = 10$ MHz   | $C_{oIF1}$                             |      | 7              |      | pF          |
| <b>Second IF preamplifier (IF 2)</b>   |  |  |      |                |      |             |
| Power gain   | Between pins 5 and 3<br>Source impedance:<br>$R_{G5} = 200 \Omega$<br>Load impedance:<br>$R_{L3} = 200 \Omega$       | $G_{IF2}$                              |      | 18             |      | dB          |
| Noise figure   |  | $NF_{IF2}$                             |      | 7              |      | dB          |
| 1 dB compression   | Pin 3  | $V_{comp}$                             |      | 500            |      | mV          |
| -3 dB cutoff frequency   | Pin 3  | $f_c$                                  |      | 50             |      | MHz         |
| Parallel input resistance  | Pin 5  | $R_{iIF2}$                             |      | 330            |      | $\Omega$    |
| Parallel input capacitance   | $f = 10$ MHz   | $C_{iIF2}$                             |      | 12             |      | pF          |
| Parallel output resistance   | Pin 3  | $R_{oIF2}$                             |      | 50             |      | k $\Omega$  |
| Parallel output capacitance  | $f = 10$ MHz   | $C_{oIF2}$                             |      | 7              |      | pF          |
| <b>Voltage regulator</b>   |  |  |      |                |      |             |
| Regulated voltage  | Pin 17   | $V_{ref}$                              |      | 3.9            |      | V           |
| Maximum output current   | Pin 17   | $I_{ref}$                              | 5    |                |      | mA          |
| Internal differential resistance,<br>$dc_{17}/di_{17}$ when $I_{17} = 0$           | Pin 17   | $r_{d17}$                              |      | 7              |      | $\Omega$    |
| Power supply suppression   | $f = 50$ Hz<br>Pin 17  | psrr                                   |      | 50             |      | dB          |

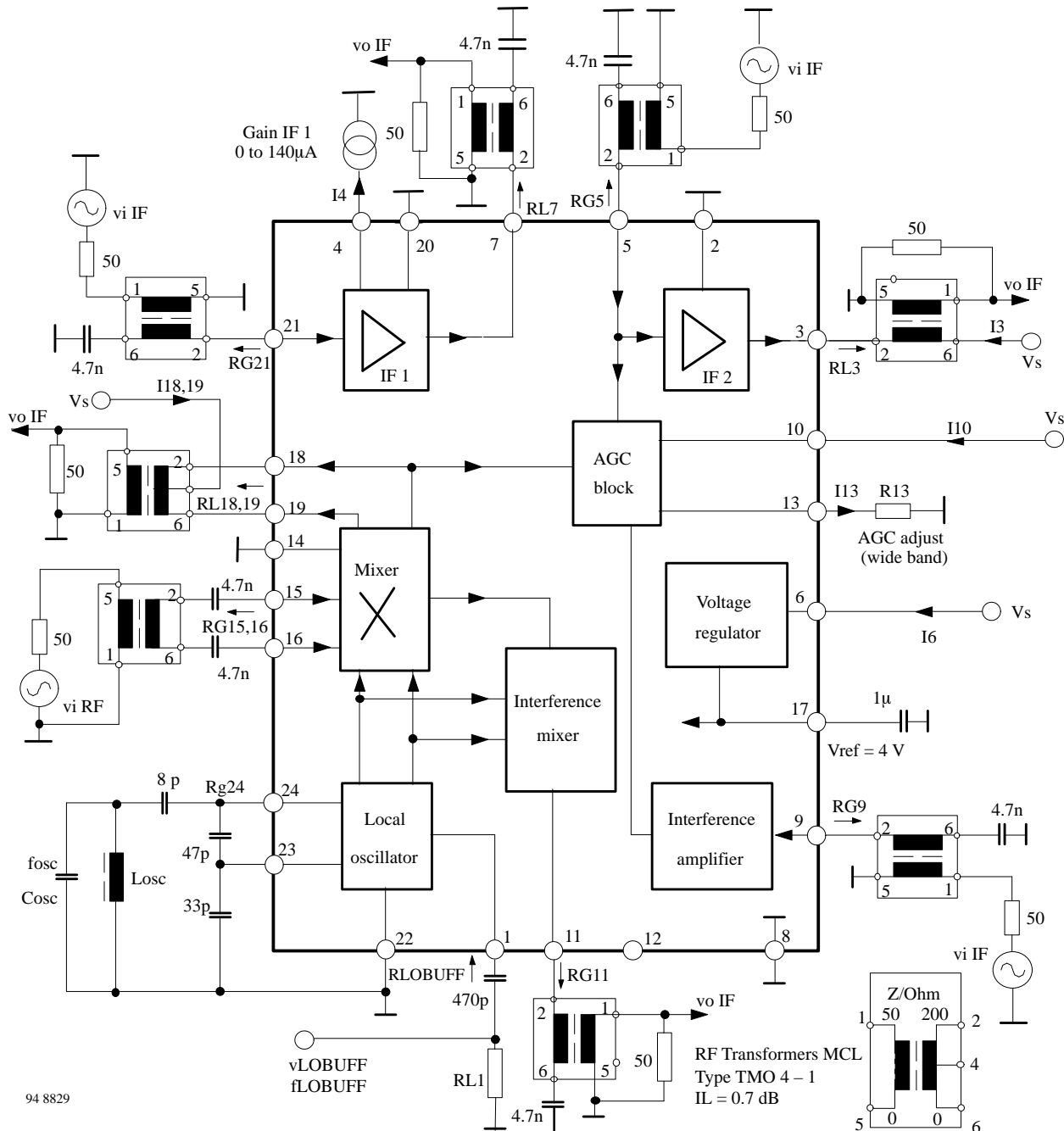
# U4065B

# TEMIC

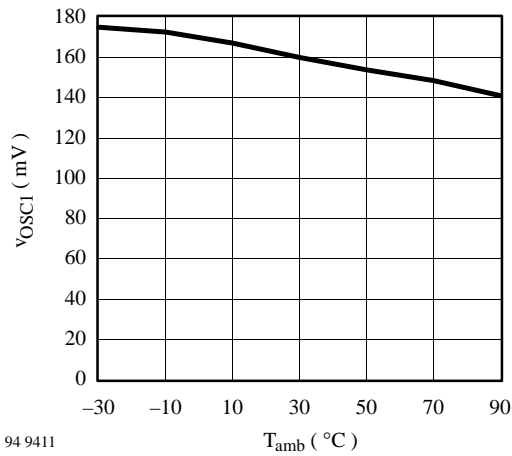
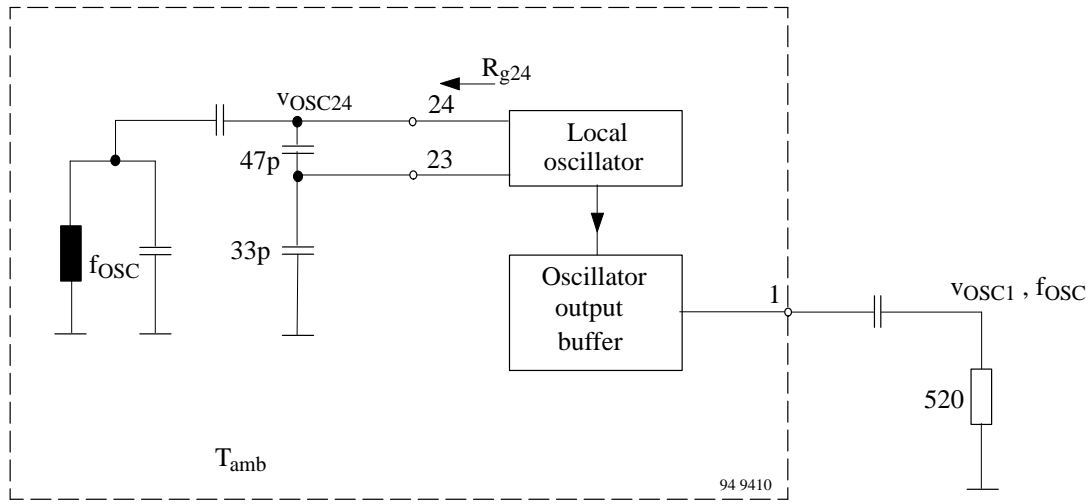
TELEFUNKEN Semiconductors

| Parameters  | Test Conditions / Pins   | Symbol      | Min. | Typ. | Max. | Unit                   |
|---|--|-------------|------|------|------|------------------------|
| <b>AGC input voltage thresholds</b> (AGC threshold current is 10 $\mu$ A at Pin 10) |  |             |      |      |      |                        |
| IF2 input   | Pin 5  | $V_{thIF2}$ |      | 86   |      | $\text{dB}\mu\text{V}$ |
| IF & detector   | Pin 9  | $V_{thIFD}$ |      | 43   |      | $\text{dB}\mu\text{V}$ |
| Mixer input level of wideband sensor  | Between Pins 15 and 16<br>$f_{iRF} = 100 \text{ MHz}$<br>$V$ at pin 13 = 0 V<br>$I$ through pin 13 = 0 A | $V_{thWB1}$ |      | 98   |      | $\text{dB}\mu\text{V}$ |
|   |  | $V_{thWB2}$ |      | 87   |      | $\text{dB}\mu\text{V}$ |

## Test Circuit



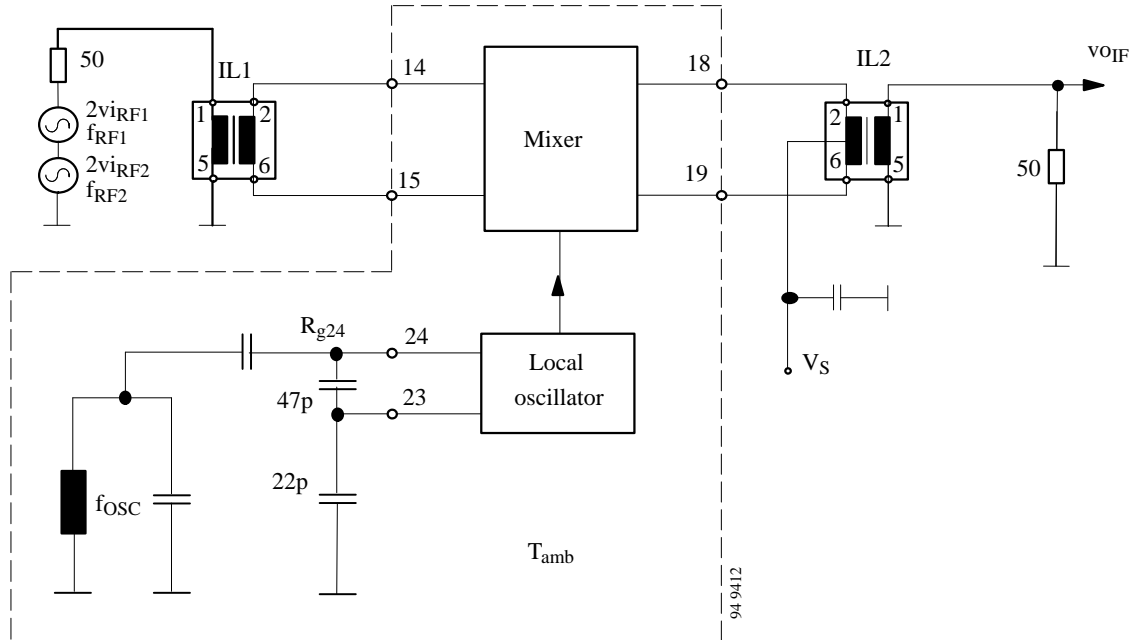
## Local Oscillator



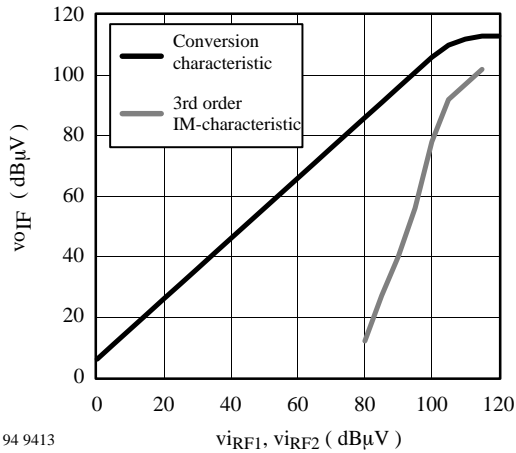
Oscillator swing versus temperature

## Mixer

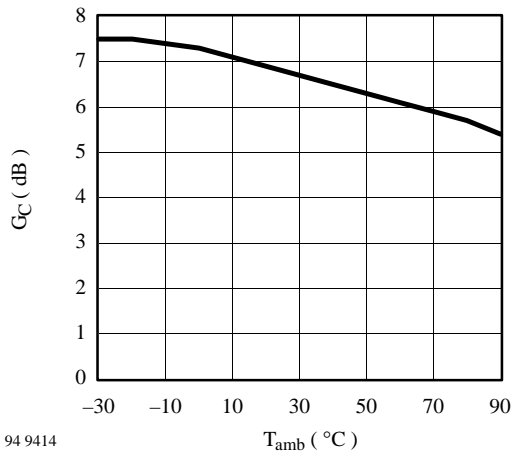
$f_{OSC} = 110.7 \text{ MHz}$ ,  $v_{OSC24} \approx 160 \text{ mV}$ ,  $f_{IF} = 10.7 \text{ MHz}$



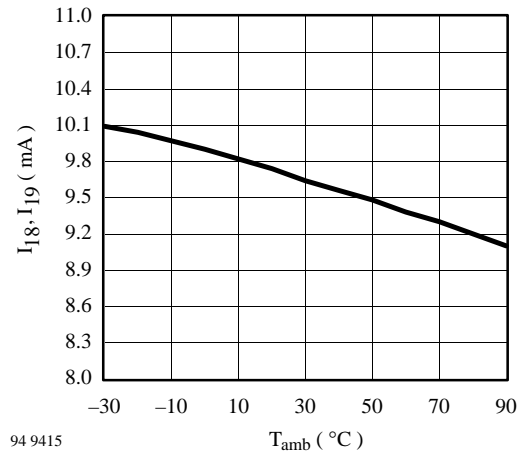
Conversion power gain  $G_C = 20 \log (v_{OIF}/v_{iRF}) + IL1 \text{ (dB)} + IL2 \text{ (dB)}$   
 IL1, IL2 insertion loss of the RF transformers



Characteristic of the mixer

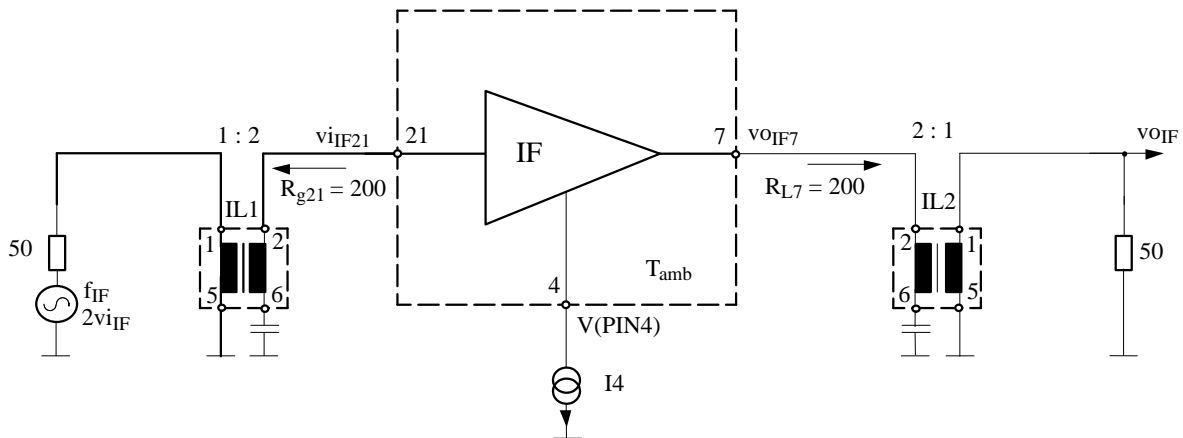


Conversion power gain of the mixer stage versus temperature



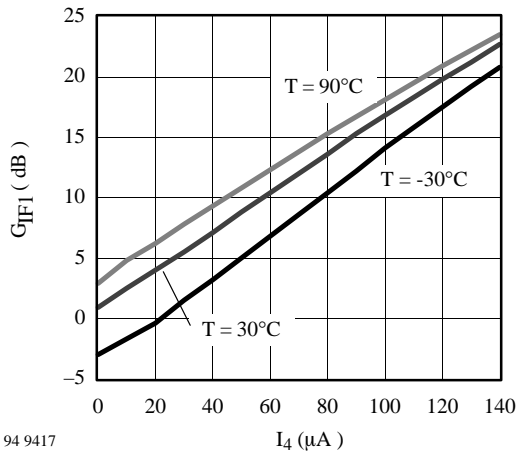
Current of the mixer stage versus temperature

## 1<sup>st</sup> IF Preamplifier

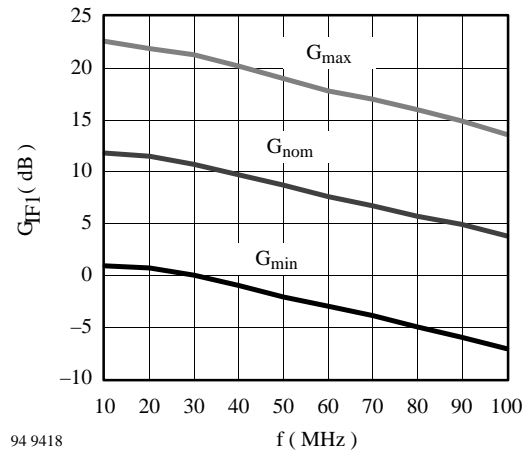


Power gain  $G_{IF} = 20 \log (v_{oIF}/v_{iIF}) + IL1 \text{ (dB)} + IL2 \text{ (dB)}$   
 IL1, IL2 = insertion loss of the RF transformers

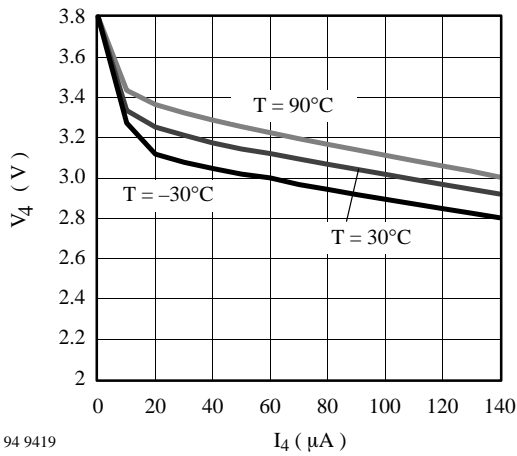
94 9416



Power gain of the first IF amplifier versus  $I_4$

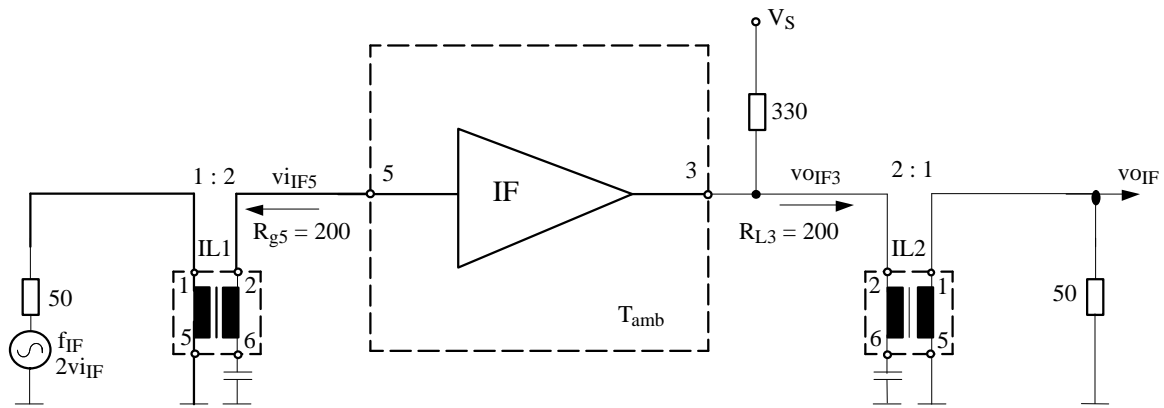


Power gain of the first IF amplifier versus frequency



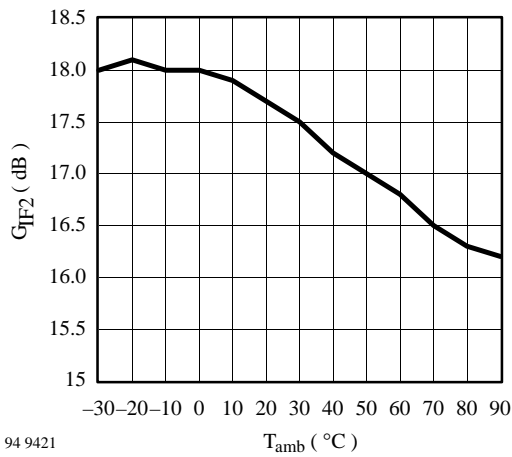
V (Pin 4) versus  $I_4$

## 2<sup>nd</sup> IF Preamplifier



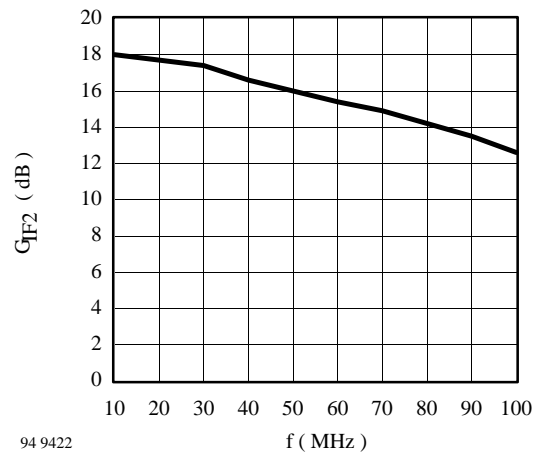
Power gain  $G_{IF} = 20 \log (v_{oIF}/v_{iIF}) + IL1 \text{ (dB)} + IL2 \text{ (dB)}$   
 IL1; IL2 = insertion loss of the RF transformers

94 9420



94 9421

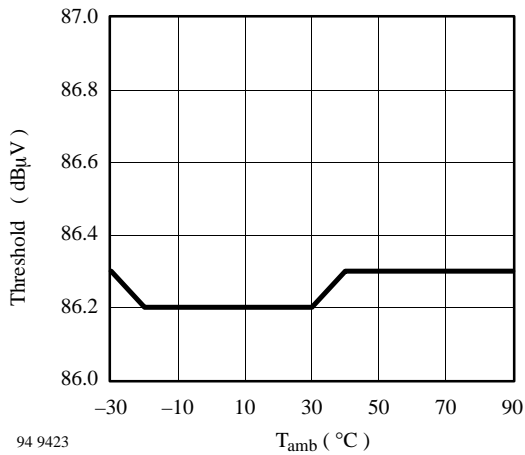
Power gain of the second IF amplifier versus temperature



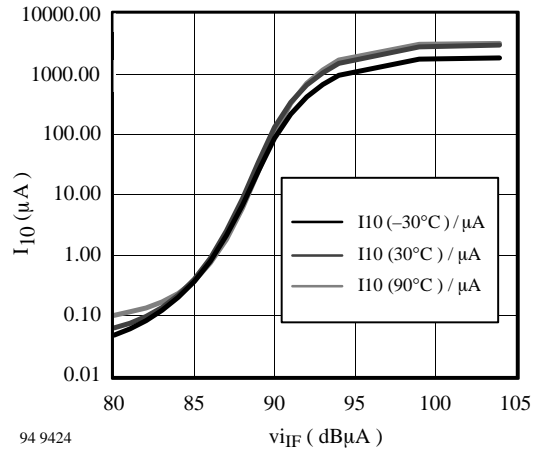
94 9422

Power gain of the second IF amplifier versus frequency

## U4065B

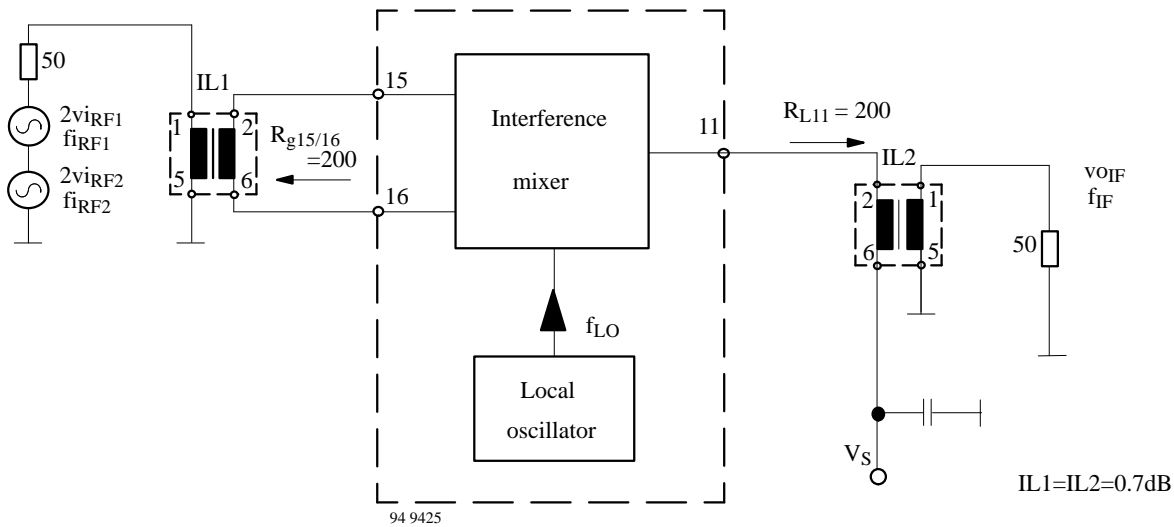


AGC threshold ( $I_{10} = 1 \mu\text{A}$ ) of the second IF amplifier versus temperature



AGC characteristic of the second IF amplifier input

### Interference Sensor (Mixer)



Test conditions for characteristic  $v_{oIF}$  versus  $v_{iRF1}$ :

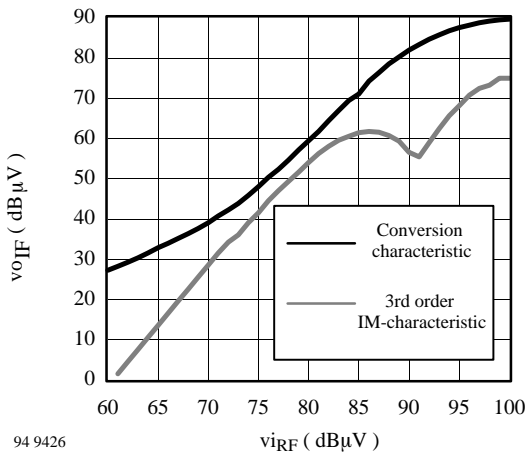
$$f_{LO} = 100 \text{ MHz}, f_{RF1} = 89.3 \text{ MHz}, v_{iRF2} = 0, f_{IF} = f_{LO} - f_{RF1} = 10.7 \text{ MHz}$$

Test conditions for 3rd order IM-characteristic  $v_{oIF}$  versus  $v_{iRF1}$ ,  $v_{iRF2}$ :

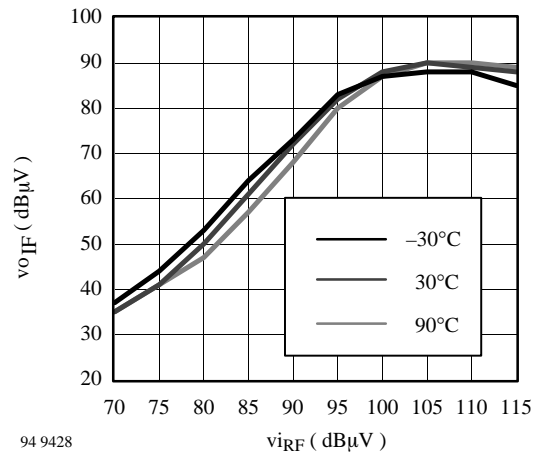
$$f_{LO} = 100 \text{ MHz}, f_{RF1} = 89.4 \text{ MHz}, f_{RF2} = 89.5 \text{ MHz}, f_{IF} = f_{LO} - (2 f_{RF1} - 1 f_{RF2}) = 10.7 \text{ MHz}$$

IL1, IL2 = insertion loss of the RF transformer

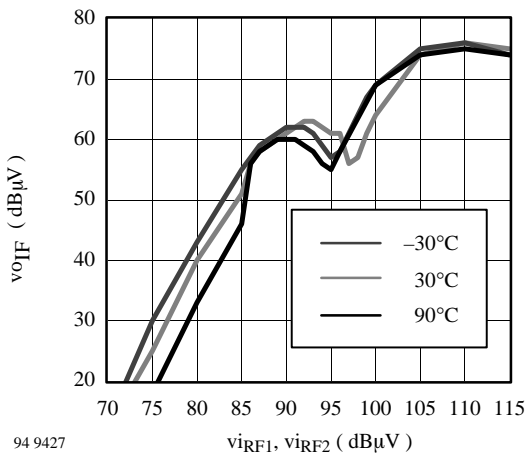




Characteristic of the interference sensor (mixer)

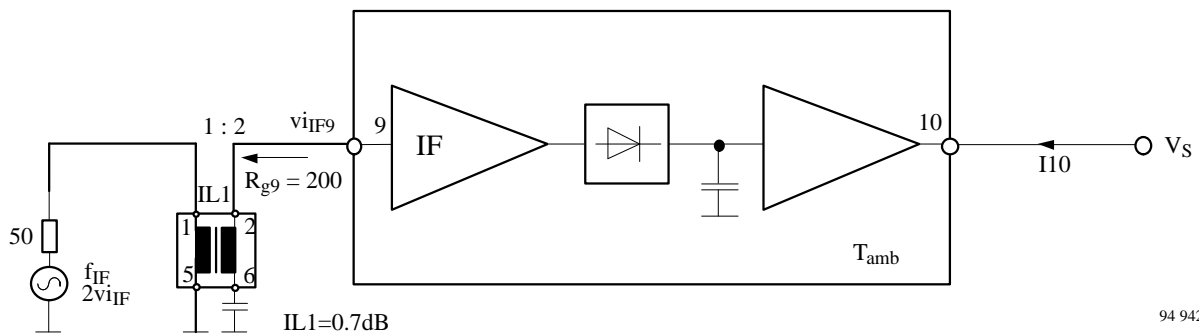


Conversion characteristic of the interference sensor (mixer)



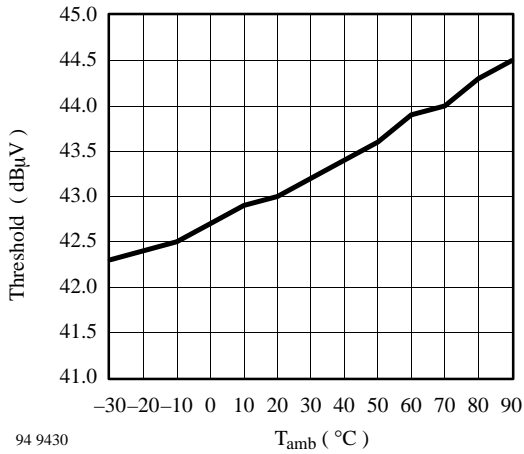
Third order interference characteristic of the interference sensor (mixer)

## Interference Sensor (Amplifier)

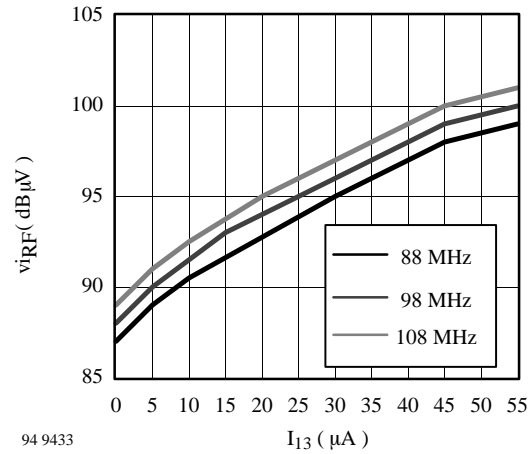


94 9429

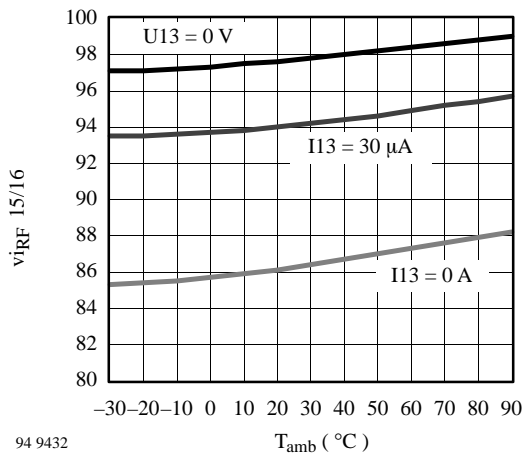
## AGC Thresholds



AGC threshold of the interference IF amplifier versus temperature

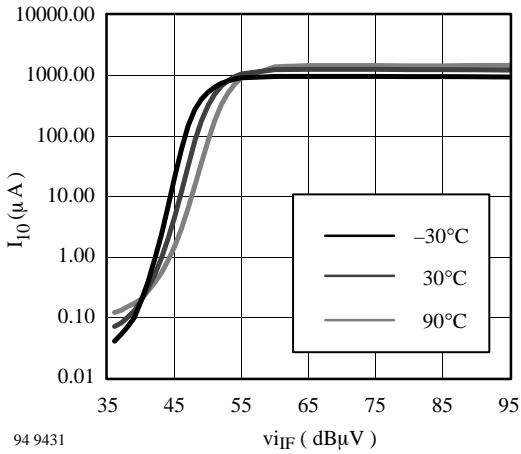


Wideband AGC threshold ( $I_{10} = 1 \mu A$ ) versus  $I_{13}$

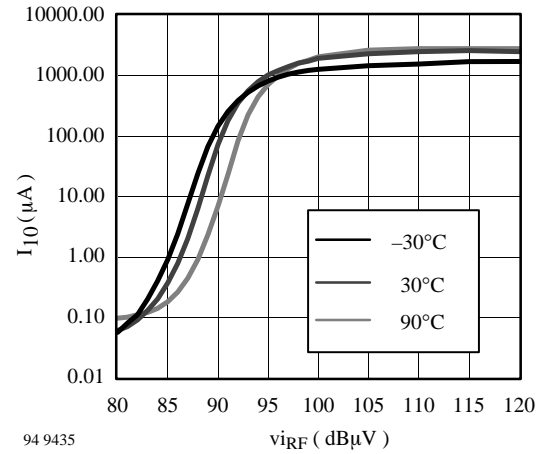


Wideband AGC threshold ( $I_{10} = 1 \mu A$ ) versus temperature

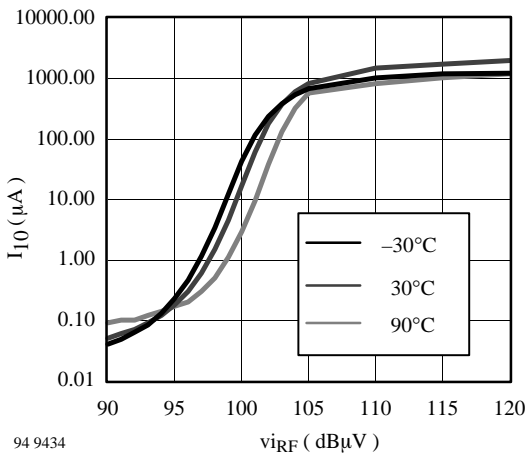
## AGC Characteristics



AGC characteristic of the interference IF & detector block

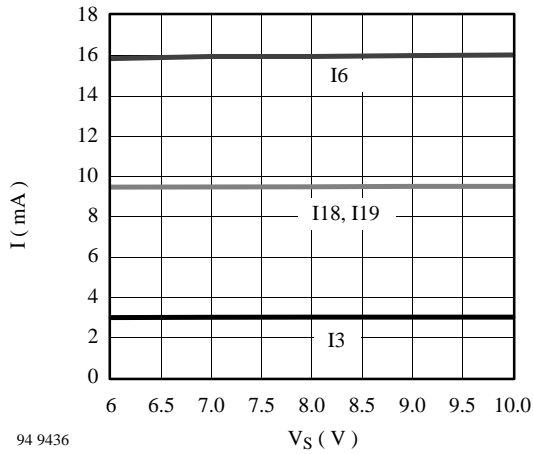


Characteristic of the wideband AGC  
(I13 = 0 V)

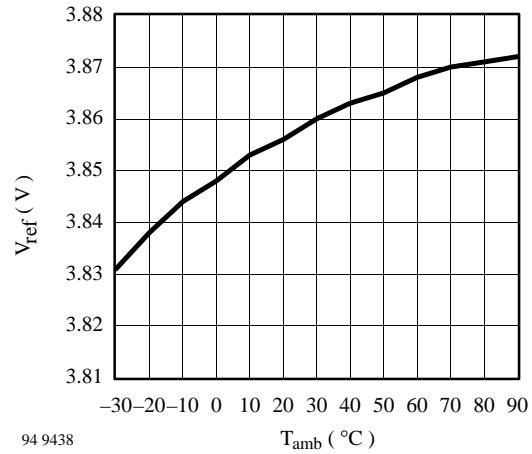


Characteristic of the wideband AGC (V13 = 0 V)

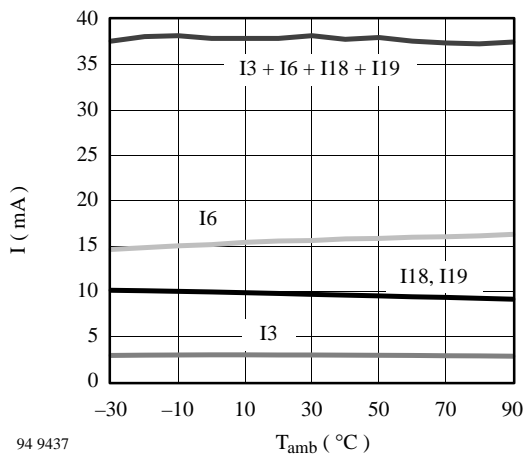
## DC Characteristics



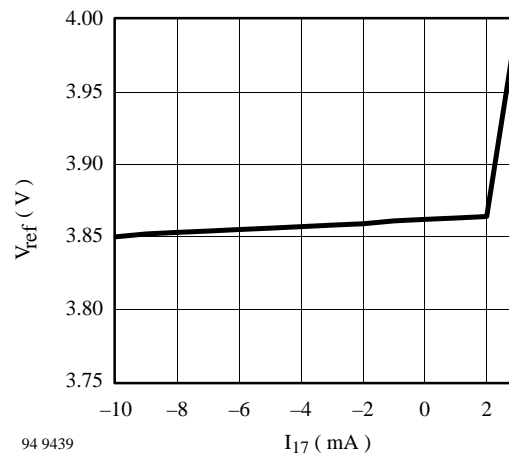
Supply currents versus supply voltage



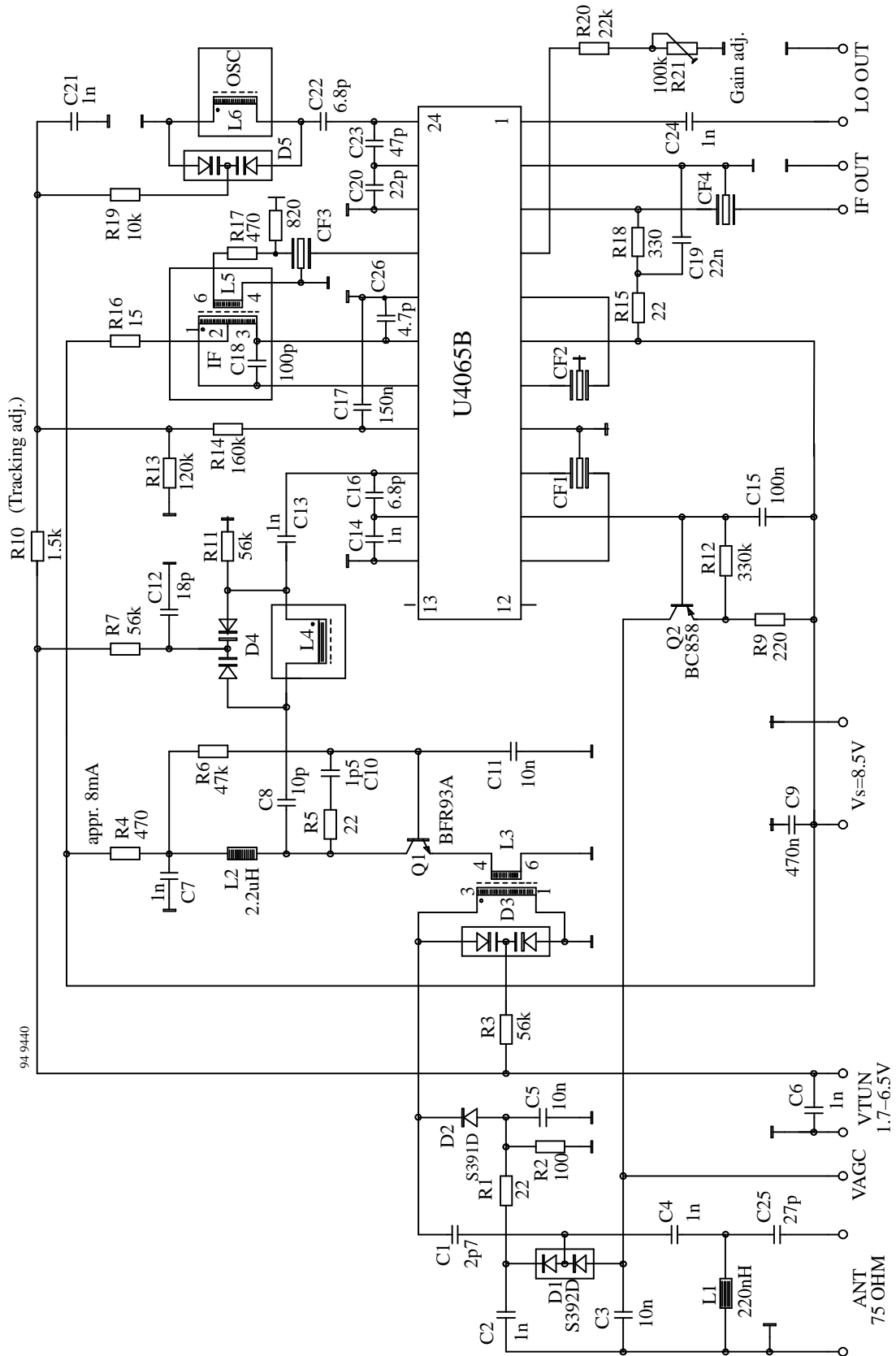
Reference voltage versus temperature



Supply currents versus temperature



Reference voltage versus  $I_{17}$



Application diagram

### Part List

| Item     | Description  |
|----------|--|
| Q1       | BFR93AR (BFR93A)                                       |
| Q2       | BC858  |
| D1       | S392D  |
| D2       | S391D  |
| D3, 4, 5 | BB804  |
| L1       | 11 turns, 0.35 mm wire, 3 mm diameter (approx. 220 nH) |
| L2       | 2.2 $\mu$ H (high Q type)                              |
| L3       | TOKO 7KL-type # 600ENF-7251x                           |

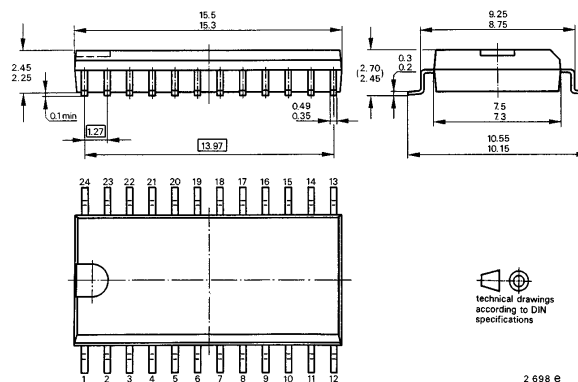
| Item      | Description                   |
|-----------|-------------------------------|
| L4        | TOKO 7KL-type # 291ENS 2341IB |
| L5        | TOKO 7KL-type # M600BCS-1397N |
| L6        | TOKO 7KL-type # 291ENS 2054IB |
| CF1       | TOKO type SKM 2 (230 KHZ)     |
| CF2, 3, 4 | TOKO type SKM 3 (180 KHZ)     |

### Ordering and Package Information

| Extended type number | Package       | Remarks                    |
|----------------------|---------------|----------------------------|
| U4065B-AFL           | SO 24 plastic |                            |
| U4065B-AFLG3         | SO 24 plastic | Taping according ICE-286-3 |

### Dimensions in mm

Package: SO 24



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## OZONE DEPLETING SUBSTANCES POLICY STATEMENT

It is the policy of **TEMIC TELEFUNKEN microelectronic GmbH** to

1. Meet all present and future national and international statutory requirements and
2. Regularly and continuously improve the performance of our products, processes, distribution and operating systems with respect to their impact on the health and safety of our employees and the public, as well as their impact on the environment.

Of particular concern is the control or elimination of releases into the atmosphere of those substances which are known as ozone depleting substances (ODSs).

The Montreal Protocol (1987) and its London Amendments (1990) will soon severely restrict the use of ODSs and forbid their use within the next ten years. Various national and international initiatives are pressing for an earlier ban on these substances.

**TEMIC TELEFUNKEN microelectronic GmbH** semiconductor division has been able to use its policy of continuous improvements to eliminate the use of any ODSs listed in the following documents.

1. Annex A, B and list of transitional substances of the Montreal Protocol and the London Amendments respectively
2. Class I and II ozone depleting substances in the Clean Air Act Amendments of 1990 by the Environmental Protection Agency (EPA) in the USA and
3. Council Decision 88/540/EEC and 91/690/EEC Annex A, B and C (transitional substances) respectively.

**TEMIC** can certify that our semiconductors are not manufactured with and do not contain ozone depleting substances.

**We reserve the right to make changes to improve technical design without further notice.**

Parameters can vary in different applications. All operating parameters must be validated for each customer application by the customer. Should the buyer use TEMIC products for any unintended or unauthorized application, the buyer shall indemnify TEMIC against all claims, costs, damages, and expenses, arising out of, directly or indirectly, any claim of personal damage, injury or death associated with such unintended or unauthorized use.

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